EECS 242: Amplifier Design Examples

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Small-Signal Transistor Model

$$
\omega_T = \frac{g_m}{C_{gs}} \qquad g_m = \frac{2I_{ds}}{V_{dsat}}
$$

$$
C_{gd} = \mu C_{gs} \qquad A_{v,0} = g_m r_o
$$

$$
r_x = \frac{1}{5g_m}
$$

$$
\overrightarrow{r_x}
$$
\n
$$
C_{\pi}
$$
\n
$$
C_{\pi}
$$
\n
$$
v_{\pi}
$$

Intrinsic transistor model

Using basic equations, let's assume we have the following process:

$$
\bullet \quad f_T = 100 \text{ GHz}
$$

$$
A_{v0} = 10, V_t = 0.3 V, V_{gs} = Vdd = 1 V, V_{dsat} = 0.7 V, I_{ds} = 5 mA
$$

$$
g_m = 14.3 \text{mA/V}, r_o = 700 \Omega, C_{gs} = 22.7 \text{ fF}
$$

$$
C_{\text{gd}} = 0.1 \, C_{\text{gs}} = 2.27 \, \text{fF}
$$

 $C_{db} = 10$ fF (arbitrary)

Simulation Setup (ADS)

• Run s-parameter StabFact simulation and plot **StabFact** various metrics StabFact1 StabFact1=stab_fact(S) Term1 $R₂$ C2. **R1** Num=1 $R1$ $C = 2.27e - 2$ R=700 Ohm $Z=50$ Ohm² 14 Ohm **VCCS** ≐ SRC1 G=1.43e-2 S GaCirole GoCircle GaCircle GpCircle S_StabCircle S_StabCircle GaCircle1 GpCircle1 GaCircle1=ga_circle(S,2,51) GpCircle1=gp_circle(S) S StabCircle UC Berkeley EECS 242 Copyright © 2009 Prof. Ali M Niknejad

S-Parameter Simulations

- Note $|S21|$ < 1 above 76 GHz, $|S12|$ is relatively small
- At low frequency, S11 and S22 dominated by capacitance

Maximum Gain Plots

Stability Factor

- Device is unconditionally stable only after 100 GHz!
- For a stable amplifier, need to add loss to input/output to make $K > 1$. The larger K, the more "safe" the design, but the less gain we can extract.

Example: 10 GHz Design

- Using this transistor, let's design an amplifier.
- Since it's bilateral, it's easier to work with the operating power gain, which is only a function of the load impedance.
- We'll notice that the source instability circle only crosses the unity Smith chart over a small region. So it's safe to assume that we can pick the optimum load and then conjugate match the input (without stability issues).
- We can always check to make sure this is true in the design.

Source/Load Stability Circles

- Since $|S11|$ and $|$ $|S22|$ <1, the origin is stable
- The source instability region is very small.
- **The load requires** careful design.

indep(S_StabCircle1) (0.000 to 51.000)
indep(L_StabCircle1) (0.000 to 51.000)

Power Gain Circles

- **Any point on a** constant power gain circle that does not lie in the instability region is a potential load.
- \blacksquare Must double check that the source stability is also satisfied for conjugate match.

indep(m1)=36
|GpCircle1=0.471 / -105.879 gain=17.011519 impedance = Z0 * (0.526 - j0.612)

indep(m2)=38
GpCircle1=0.426 / -72.669 gain=18.011519 impedance = Z0 * (0.882 - j0.877)

 $m3$ indep(m3)=33
GpCircle1=0.404 / -175.145 gain=20.011519 $\frac{3}{2}$ impedance = Z0 * (0.425 - j0.035)

Example Amplifier Design

- **Design a single stage amplifier with** $G = 14$ **dB and input match.**
- Since input must be matched, let's select the load for gain/stability.
- $MSG = 20$ dB; back-off by 6 dB and plot $Gp = 14$ dB circles. Also plot load stability.
- Select load far away from instability region.

Output Load Circles

• Point m1 is far away from instability region for load.

Output Matching Network

Output Load Match

Note that the

Gp=14dB circle

crosses origin

Origin still in

due to input

 \blacksquare Since S11 is

match

stable region.

Power gain is low

mismatch (-4 dB)

"stable", design

input matching

network for imp

Input Match

Complete Amp Schematic

- Out match is simple L matching network; component values are reasonable.
- Other possibilities include T-line/stub matching.
- Input match component values are not very practical but okay for demonstration.

Source Matching

- Input is matched and now the realized power gain is 14 dB
- The source stability circle, though, is dangerously close to the origin.

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Match/Gain Plots

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Comments on Design

- Input match looks good $(S11 < 0.11)$
- Power gain is as desired: $G = 14$ dB
- Output match is terrible $(S22 \sim 1)$ and close to instability!
- **We see that this design requires some iteration to arrive at** the best input/output match pair to give gain and good stability.
- In practice a lossy matching network (or added loss) feedback to device) helps to create a good match.