EECS 242: Amplifier Design Examples

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Small-Signal Transistor Model

$$\omega_T = \frac{g_m}{C_{gs}} \qquad g_m = \frac{2I_{ds}}{V_{dsat}}$$
$$C_{gd} = \mu C_{gs} \qquad A_{v,0} = g_m r_o$$
$$r_x = \frac{1}{5g_m}$$

$$\begin{array}{c}
 & C_{\mu} \\
\hline
 & r_x \\
 & r_x \\$$

Intrinsic transistor model

Using basic equations, let's assume we have the following process:

•
$$f_T = 100 \text{ GHz}$$

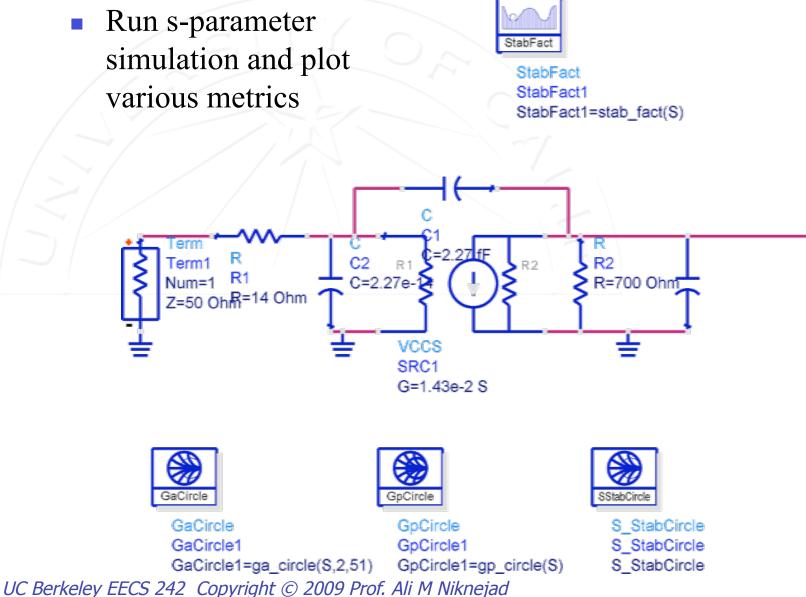
•
$$A_{v0} = 10, V_t = 0.3V, V_{gs} = Vdd = 1V, V_{dsat} = 0.7V, I_{ds} = 5mA$$

•
$$g_m = 14.3 \text{mA/V}, r_o = 700\Omega, C_{gs} = 22.7 \text{ fF}$$

•
$$C_{gd} = 0.1 C_{gs} = 2.27 \text{ fF}$$

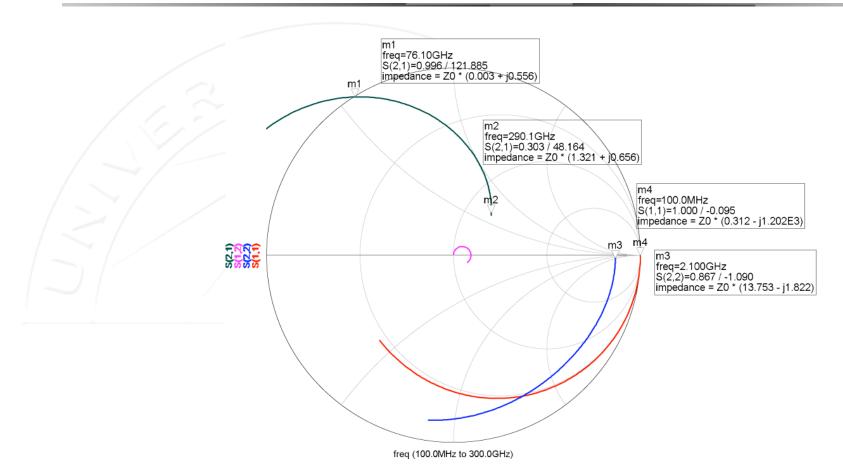
• $C_{db} = 10 \text{fF} \text{ (arbitrary)}$

Simulation Setup (ADS)



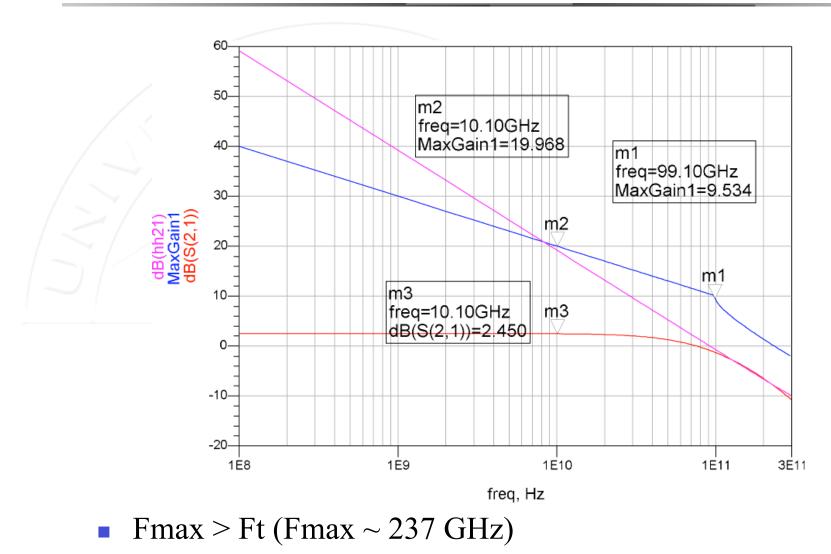
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S-Parameter Simulations

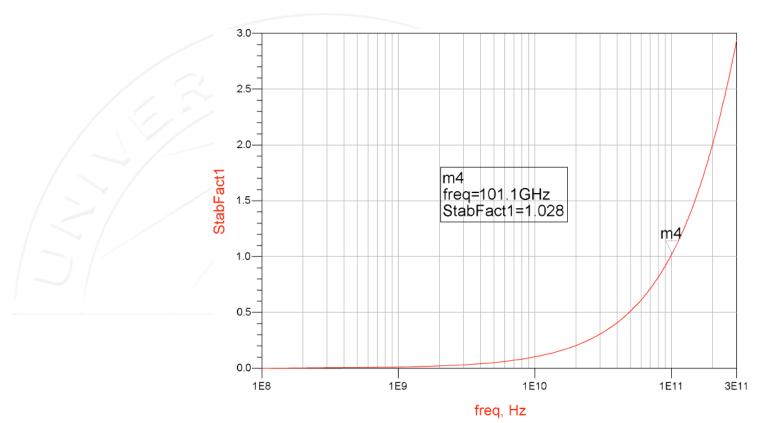


- Note |S21| < 1 above 76 GHz, |S12| is relatively small
- At low frequency, S11 and S22 dominated by capacitance

Maximum Gain Plots



Stability Factor



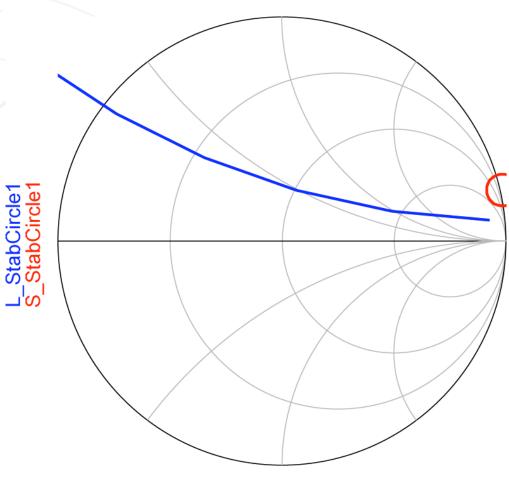
- Device is unconditionally stable only after 100 GHz!
- For a stable amplifier, need to add loss to input/output to make K > 1. The larger K, the more "safe" the design, but the less gain we can extract.

Example: 10 GHz Design

- Using this transistor, let's design an amplifier.
- Since it's bilateral, it's easier to work with the operating power gain, which is only a function of the load impedance.
- We'll notice that the source instability circle only crosses the unity Smith chart over a small region. So it's safe to assume that we can pick the optimum load and then conjugate match the input (without stability issues).
- We can always check to make sure this is true in the design.

Source/Load Stability Circles

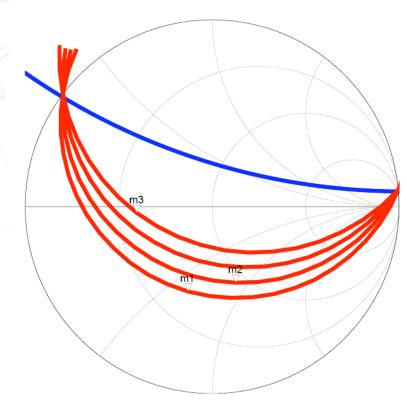
- Since |S11|<1 and |
 S22|<1, the origin is stable
- The source instability region is very small.
- The load requires careful design.



indep(S_StabCircle1) (0.000 to 51.000) indep(L_StabCircle1) (0.000 to 51.000)

Power Gain Circles

- Any point on a constant power gain circle that does not lie in the instability region is a potential load.
- Must double check that the source stability is also satisfied for conjugate match.

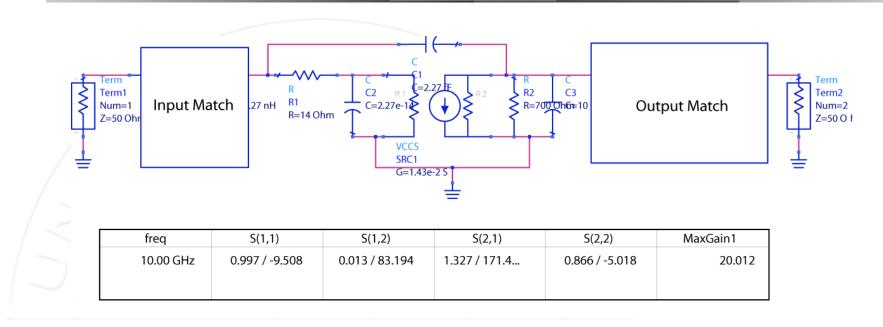


m1 indep(m1)=36 GpCircle1=0.471 / -105.879 gain=17.011519 impedance = Z0 * (0.526 - j0.612)

m2 indep(m2)=38 GpCircle1=0.426 / -72.669 gain=18.011519 impedance = Z0 * (0.882 - j0.877)

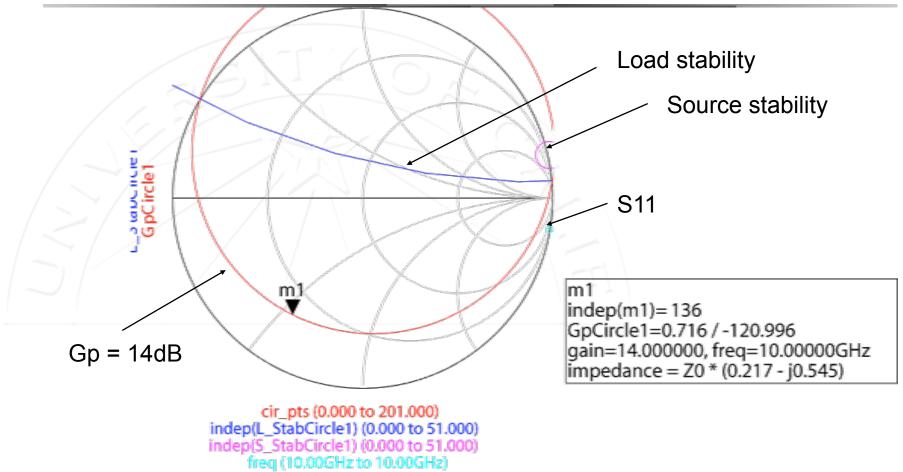
m3 indep(m3)=33 GpCircle1=0.404 / -175.145 gain=20.011519 impedance = Z0 * (0.425 - j0.035)

Example Amplifier Design



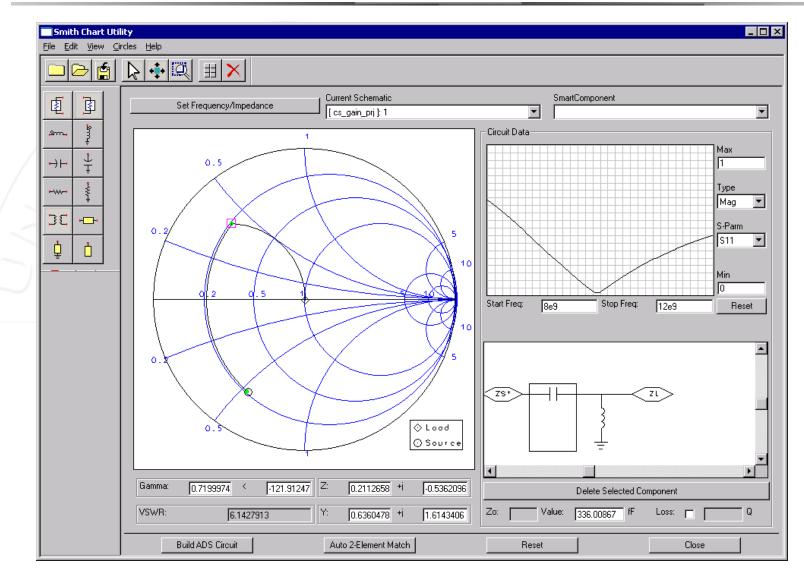
- Design a single stage amplifier with G = 14 dB and input match.
- Since input must be matched, let's select the load for gain/stability.
- MSG = 20 dB; back-off by 6 dB and plot Gp = 14 dB circles. Also plot load stability.
- Select load far away from instability region.

Output Load Circles

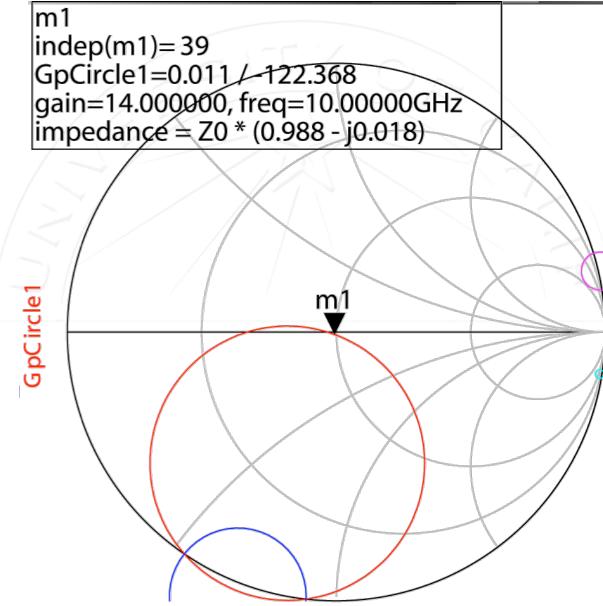


• Point m1 is far away from instability region for load.

Output Matching Network

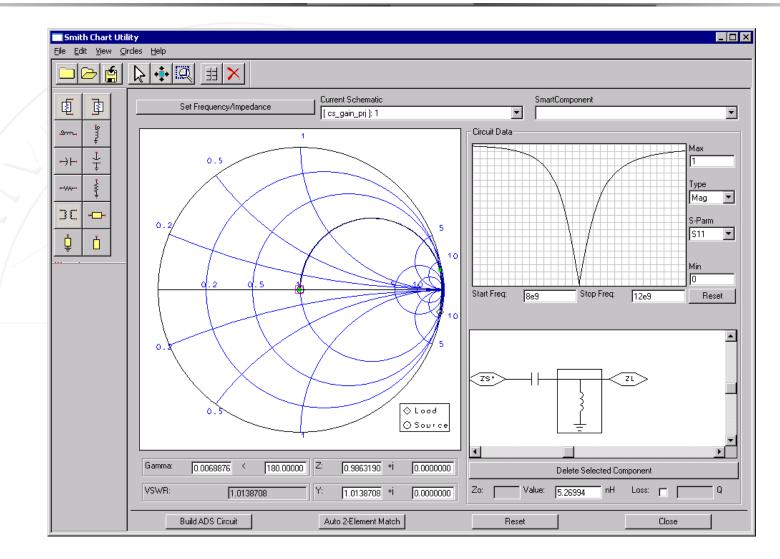


Output Load Match

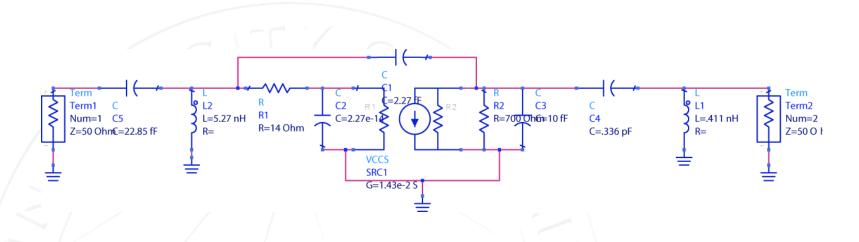


- Note that the Gp=14dB circle crosses origin
- Origin still in stable region.
- Power gain is low due to input mismatch (-4 dB)
- Since S11 is "stable", design input matching network for imp match

Input Match

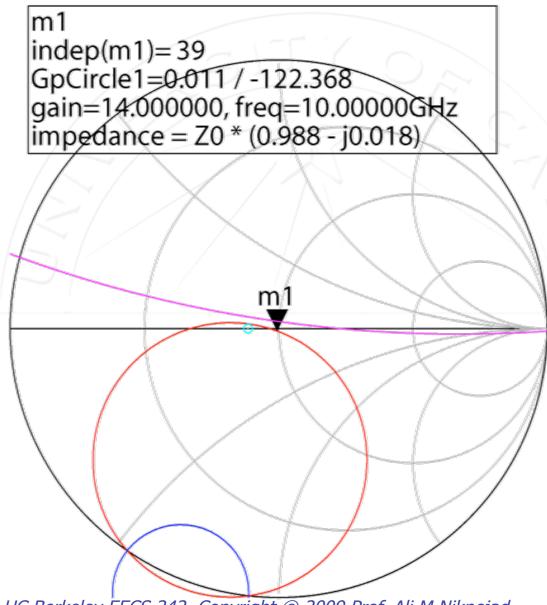


Complete Amp Schematic



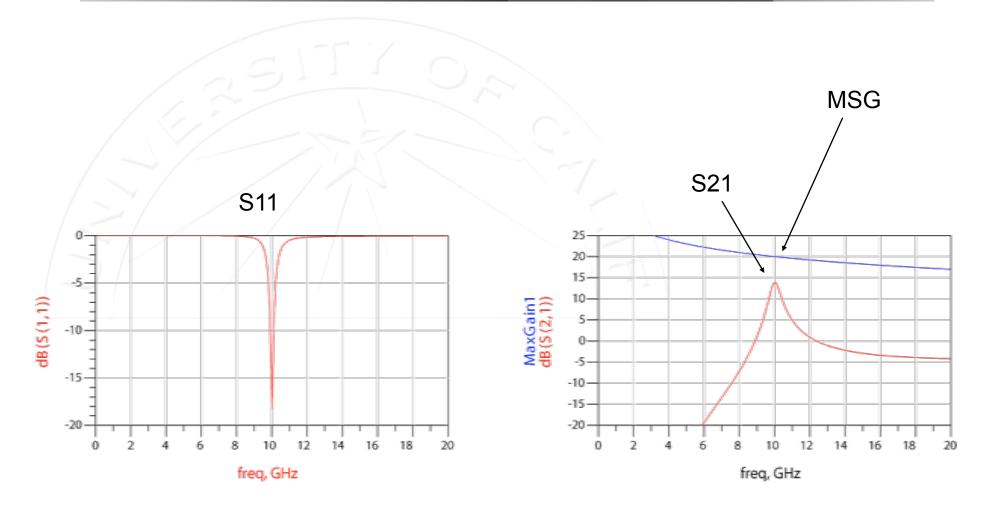
- Out match is simple L matching network; component values are reasonable.
- Other possibilities include T-line/stub matching.
- Input match component values are not very practical but okay for demonstration.

Source Matching



- Input is matched and now the realized power gain is 14 dB
- The source stability circle, though, is dangerously close to the origin.

Match/Gain Plots



Comments on Design

- Input match looks good (S11 < 0.11)
- Power gain is as desired: G = 14 dB
- Output match is terrible (S22 \sim 1) and close to instability!
- We see that this design requires some iteration to arrive at the best input/output match pair to give gain and good stability.
- In practice a lossy matching network (or added loss/ feedback to device) helps to create a good match.